

Least-Squares PDV Error Estimation

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Abstract

A new method for analyzing PDV data is discussed. It is based on using preliminary frequency estimates garnered from a windowed spectrogram and an interpolated FFT algorithm. A novel third step produces a least squares estimate with mathematically precise error bounds that meet the Cramer-Rao least variance lower bound. This approach can be used to study phenomena at the sub-nanosecond scale by allowing the window length to change adaptively to account for signal speed and noise level, while minimizing the expected errors.

Principal authors of the software

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Thanks to:

David Holtkamp at LANL who supplied several utility-routines that were used in testing the software.

Ted Strand at LLNL who supplied some very helpful and interesting example data to “stress test” these algorithms.

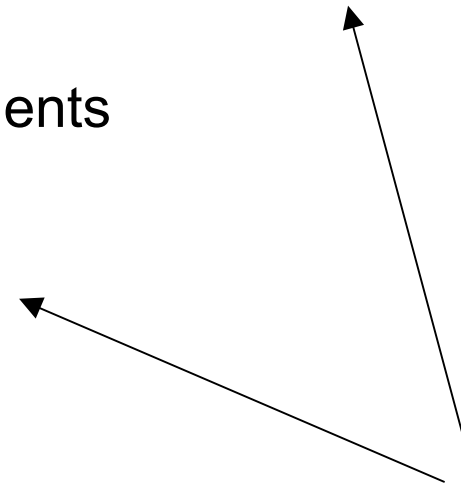
Overview of PDV Analysis Algorithm

- 1) Filter raw data to remove time varying offset.
- 2) Estimate noise level from 1st 1000 points.
- 3) Perform spectrogram of length $N=2^p$ with cosine window
- 4) Find peaks (old algorithm)
- 5) Refine peaks with interpolated FFT algorithm
- 6) Refine further with least squares
- 7) Derive error estimates & show data with error bars
- 8) *Adaptively vary interval length to lower error*

} Under construction

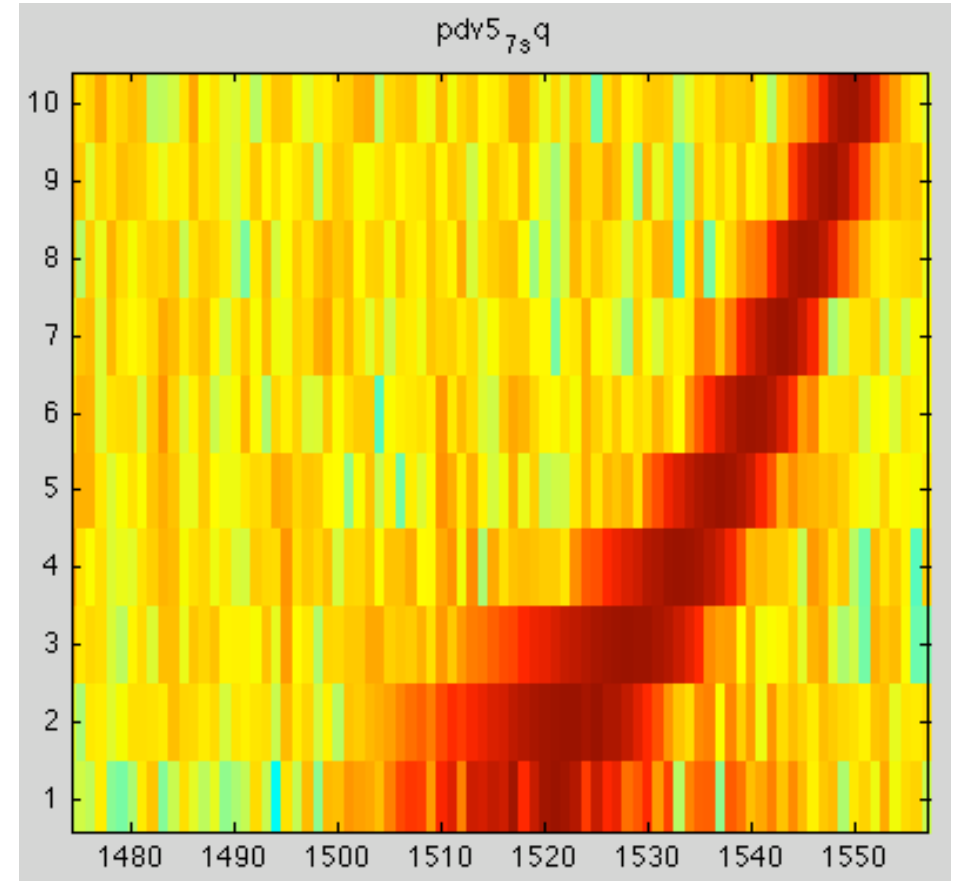
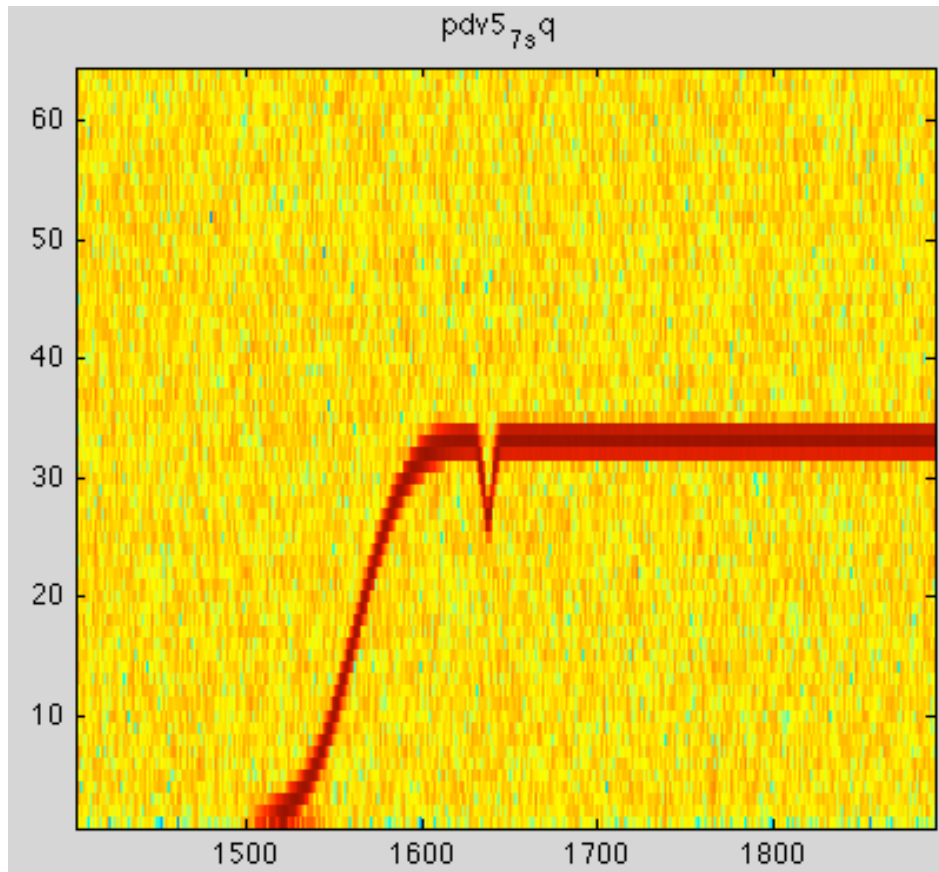
Generating the Spectrogram

- For each selected time interval of length N
 - Multiply measured signal by cosine window.
 - Calculate magnitudes of Fourier coefficients of result.
- Perform for intervals centered at $(1/2)N, N, (3/2)N, \dots$
- Result is a vector of Fourier coefficients
 - Length of each vector in $N/2$
 - Time spacing of vectors is $(N/2)\Delta t$
- Plot and save result



In next version, this has changed to a time spacing/overlap of $(1/8)N \Delta t$

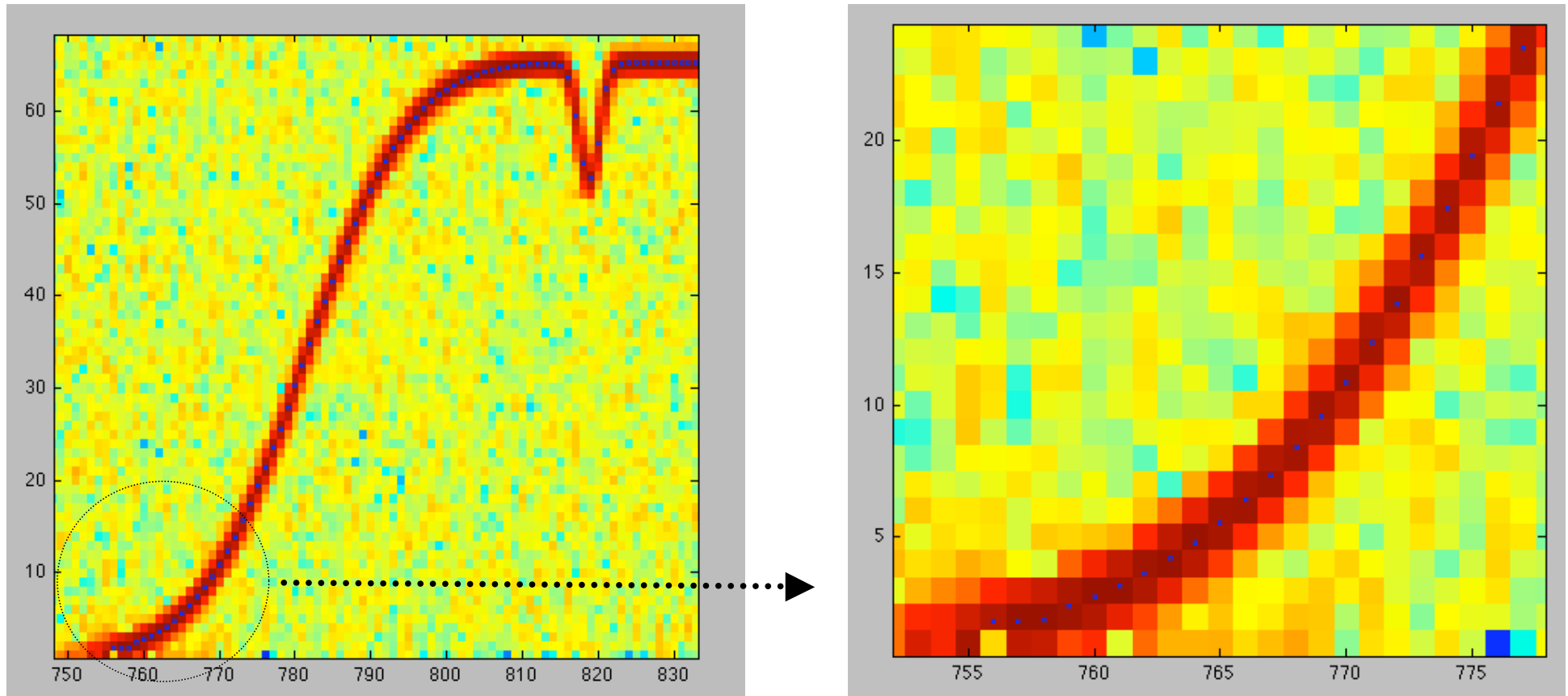
Example Spectrograms



$N = 128$ for 64 vertical slices

Horizontal slices are
separated by $64\Delta t$.

Find and Refine Peaks



- For each vertical column, find pixel with maximum intensity.
- Interpolated Fast Fourier-Transform (IpFFT) uses neighboring intensity value to give fractional pixel value.

Least-Squares Estimation

- Uses original signal as a function of time.
- One frequency estimate for each spectrogram estimate—using the same interval of data.
- Interpolated FFT estimate (based on the spectrogram) is the starting value for an iterative non-linear least squares routine.

Fitting the data to the model

$$y(t) = A(t + \delta t) \sin(X(t + \delta t)) + \eta(t)$$

$$\text{for } t \in \left[t_c - \frac{N\Delta t}{2}, t_c + \frac{N\Delta t}{2} \right]$$

with

$$X(t) = x_0 + x_1(t - t_c) + x_2(t - t_c)^2,$$

$$A(t) = A_0 + A_1(t - t_c),$$

$$\eta(t) = \text{noise}, \quad \delta t = \text{time jitter},$$

$$\text{and } f(t_c) = \frac{x_1}{2\pi} \text{ the frequency estimate.}$$

Least-Squares Estimation

$$\text{Fit } y(t) = A(t)\sin(X(t))$$

with

$$X(t) = x_0 + x_1(t - t_c) + x_2(t - t_c)^2,$$

$$A(t) = A_0 + A_1(t - t_c).$$

- Spectrogram gives initial estimate for x_0 . The spectrogram can also provide estimates for x_1 , x_2 , and A_0 .
- Use linear least squares to yield an initial estimate for the *remaining parameters*.
- Use those starting values to begin the nonlinear least squares routine.

Error Estimates

Key assumptions are that the true signal has two continuous derivatives and that the oscilloscope noise is white.

Fitting error, due to
signal not fitting the
model

$$E_{2\sigma} \cong \sqrt{c_1^2 \sigma_{noise}^2 + c_2^2 \ddot{v}(t)^2 + c_3^2 \ddot{A}(t)^2}$$

Random error, due
to oscilloscope
noise and time-jitter.

Random & Fitting Error Estimates

$$E_{2\sigma} \cong \sqrt{c_1^2 \sigma_{noise}^2 + c_2^2 \dot{v}(t)^2 + c_3^2 \ddot{A}(t)^2}$$

Random error, due to oscilloscope noise and time jitter.

Fitting error, 2nd derivative estimated with divided difference of velocity

- Random error based on oscilloscope noise being white (from tests of oscilloscopes.) The term σ_{noise} is the standard deviation of the random effects that include oscilloscope noise and time-jitter.

Error Estimates - cont'd

$$E_{2\sigma} \cong \sqrt{c_1^2 \sigma_{noise}^2 + c_2^2 \ddot{v}(t)^2 + \dots}$$

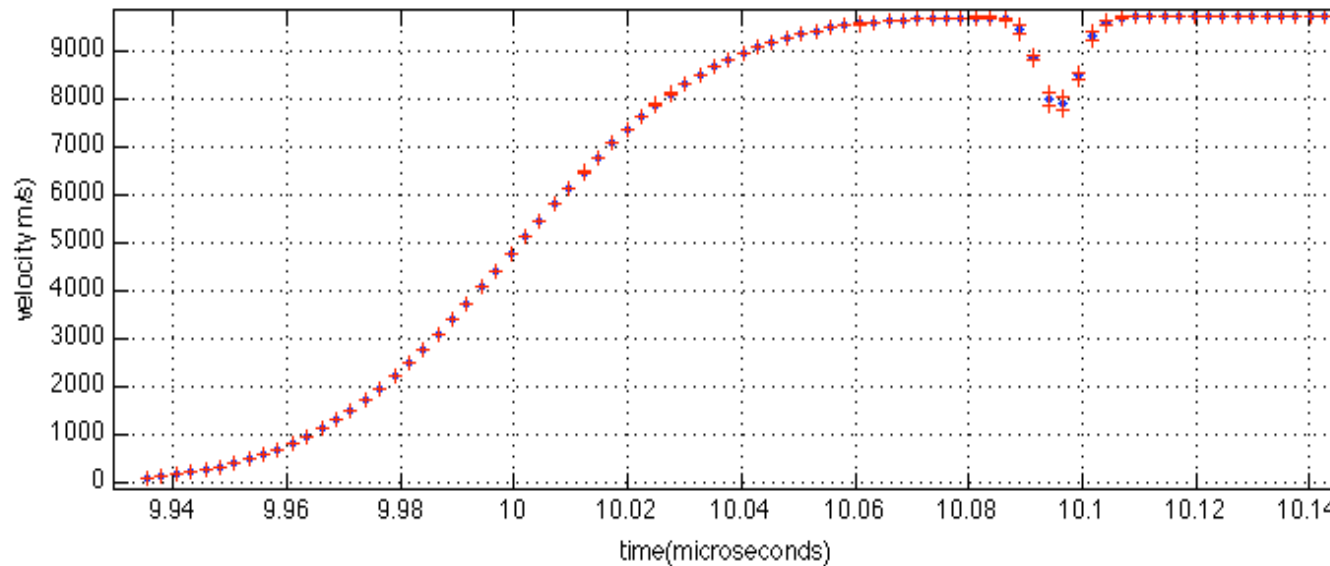
Note that

$$c_1 \cong \frac{C_1 \Delta t^{-3/2} N^{-3/2}}{A_0}, \text{ and } c_2 \cong C_2 \Delta t^2 N^2.$$

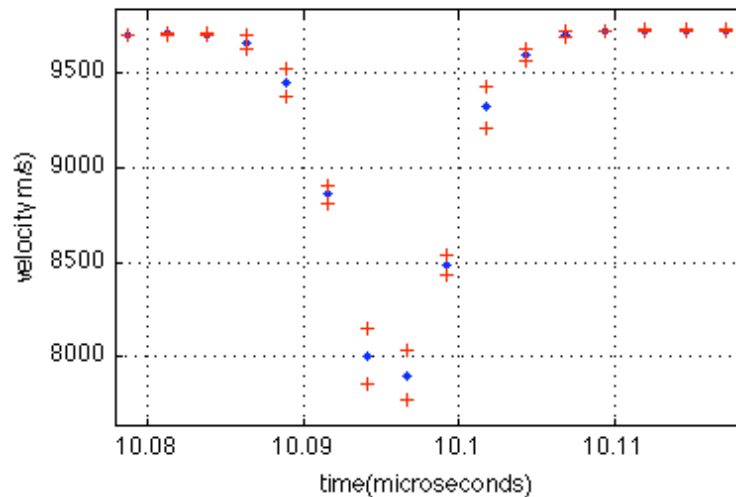
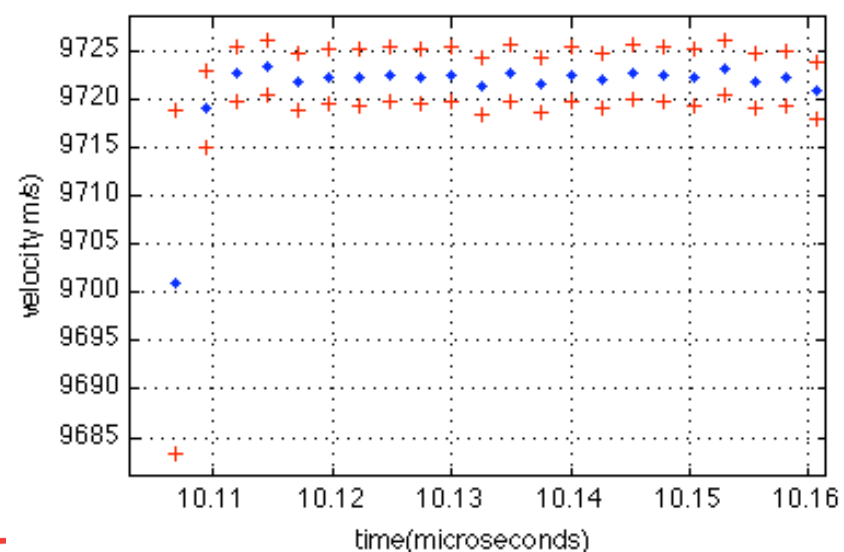
These approximations can be useful to know but are not used by the analysis software.

Some examples ... well ... okay ... 1 example

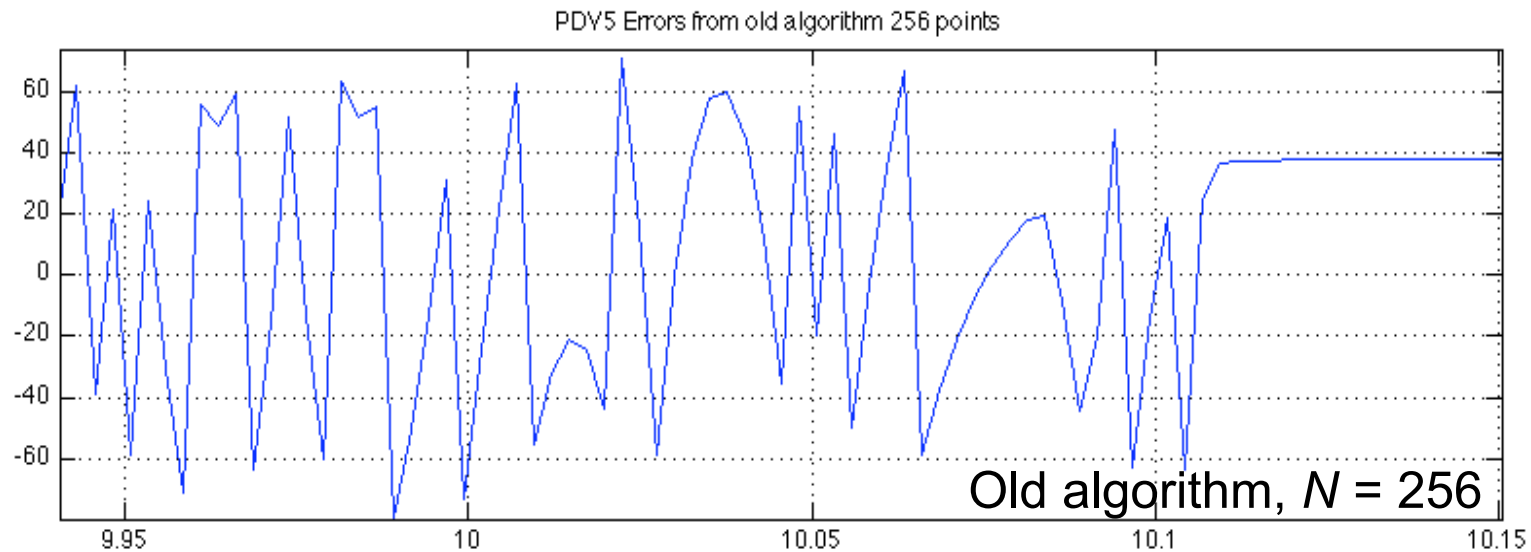
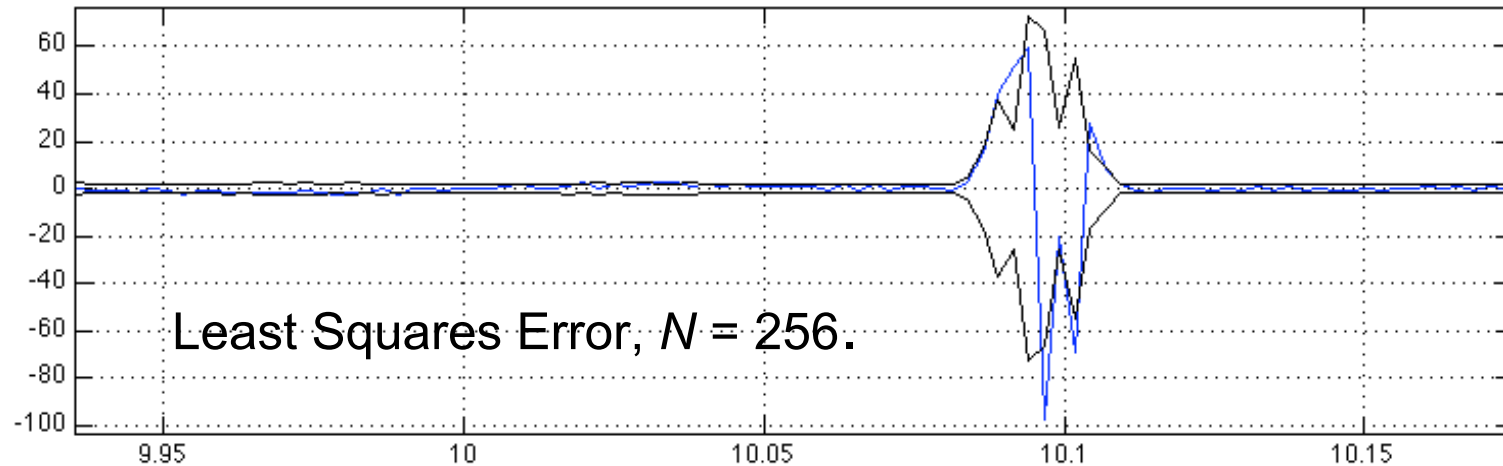
Example Least-Squares Output

velocity and error from $\text{pdv}_{\text{BS}}^5 q$ 

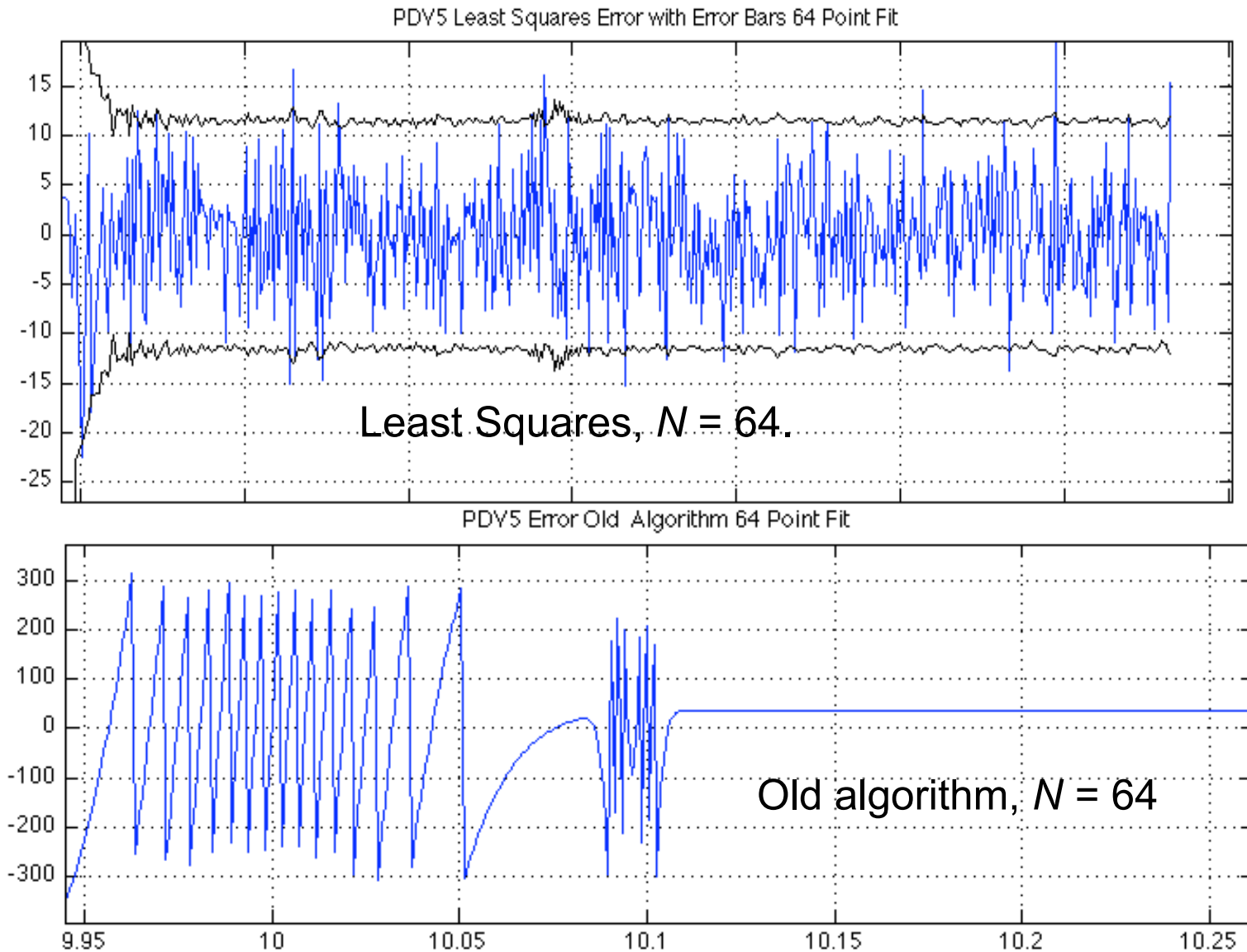
*Subinterval of
 $N = 256$ pts*

velocity and error from $\text{pdv}_{\text{BS}}^5 q$ velocity and error from $\text{pdv}_{\text{BS}}^5 q$ 

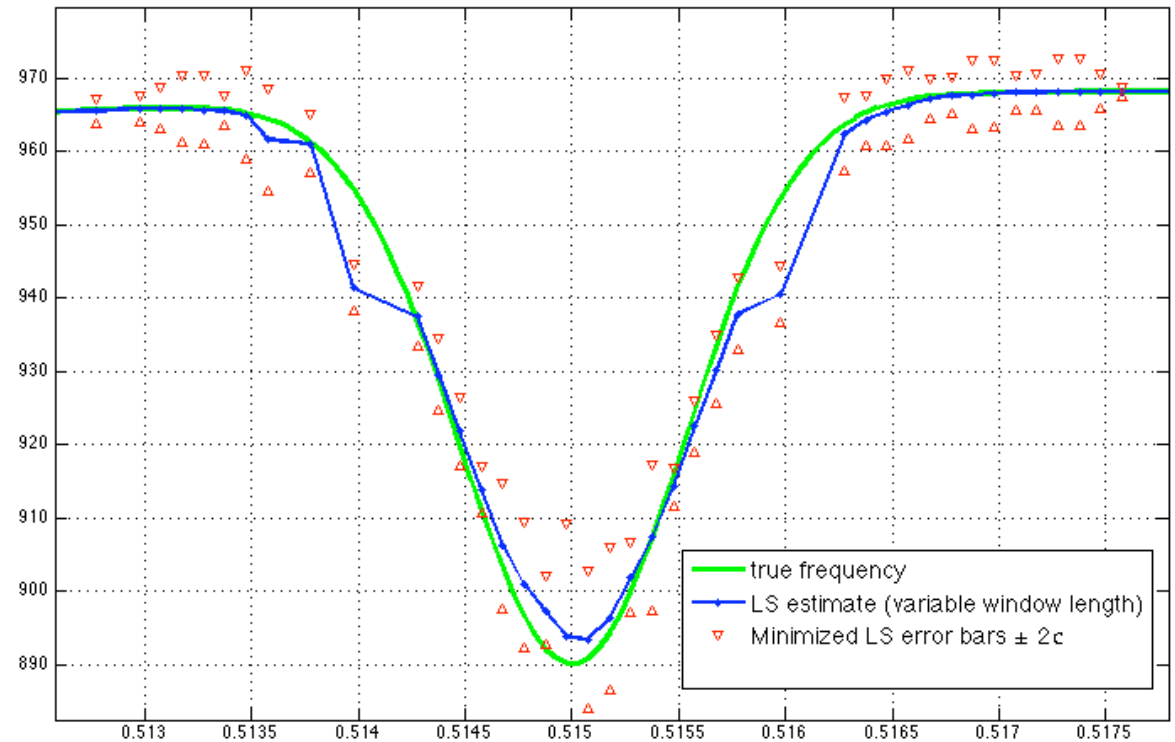
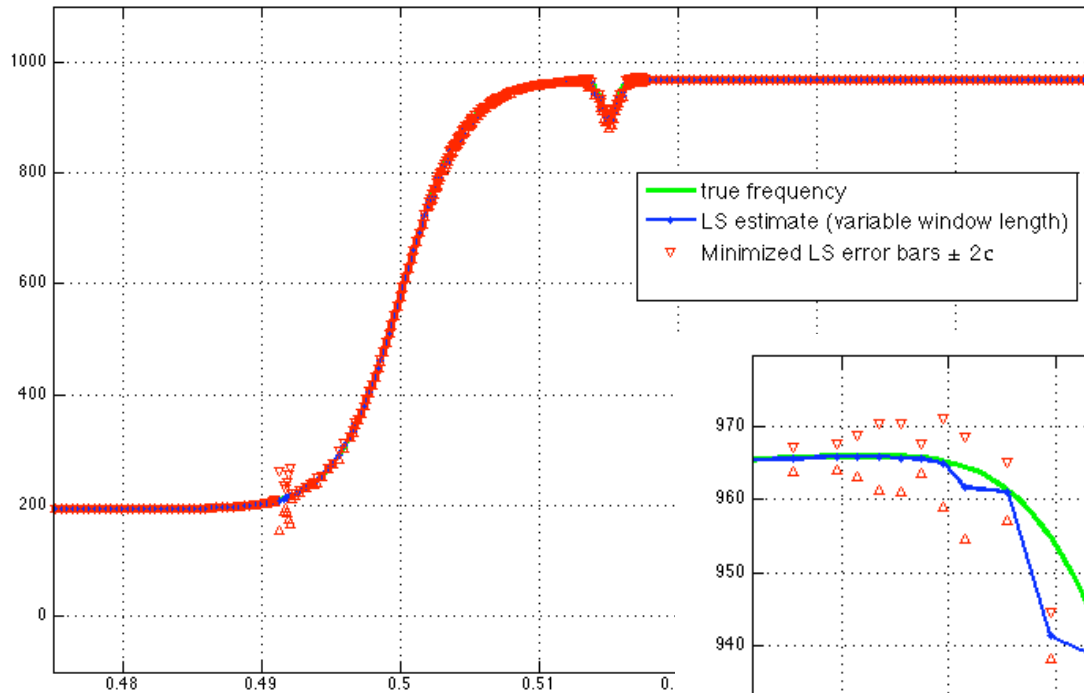
Comparison of Errors



Shorter Window Size



A preview of upcoming attractions ...



Summary

- New GUI-based software produces error bars that account for measurement noise, time-jitter, and fitting errors.
- Method produces the best, least-squares, unbiased estimate of the frequency. Mathematically impossible to do better on average for a given window length.
- Adaptively changing the window length, even more accurate error bars can be calculated. Determining the best window length to capture the phenomena of interest will no longer be an issue.

Issues to fix ... for adaptive windowing

- Non-linear least-squares routines can often find “wrong” solution
 - These routines are very sensitive to the initial IpFFT-based estimate.
 - Constrained minimization routines commercially are available.
 - Use residual sizes and error-bar length to “toss out” bad estimates.
- Computational time for larger data sets can be quite long. In testing, one example took 2 hrs to complete.
 - It’s anticipated that this can be sped up considerably.
 - Fixing the window length (however) is much faster.